Investigation of the Nonlinear Electromagnetic Energy Harvesters From Hand Shaking

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Abstract—This paper presents the design, modeling, and optimization of an electromagnetic energy harvesting (EMEH) device with various tube length and winding coil width. The nonlinear magnetic-spring configuration is employed for generating sufficient power from hand shaking of irregular and low-frequency vibrations. Based on the modeling and simulation, longer tube length of the EMEH device results in lower resonant frequency and stronger nonlinearity of the system and thus is more efficient for low-frequency harvesting. From the hand shaking test, it is found that a longer tube length and a shorter winding coil width could induce a higher power generation. The optimized device of tube length of 66 mm and winding coil width of 10 mm has achieved maximum power outputs of 568.66 μ W at the hand shaking acceleration of 1 g and frequency of 6.7 Hz, which corresponds to the power density of 90.67 μ W/cm³. Further improvement of the device has been achieved by increasing the coil length to 40 m. This device provides maximum power outputs of 825.36 μ W at the hand shaking acceleration of 1.56 g and frequency of 6.7 Hz. This paper demonstrates a feasible design of nonlinear EMEH device with impressive output performance from hand shaking.

Index Terms—Electromagnetic energy harvester (EMEH), nonlinear, magnetic-spring, hand shaking.

I. INTRODUCTION

THE concept of vibrational energy harvesting has gained much attention in recent years because of the rapid development on microelectronic devices and systems with low power consumption. A particular motivating scenario is the self-powered wireless sensor network for body and structure health monitoring [1]–[4]. The requirements for an energy

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harvester are high power efficiency, high durability, and wide operating frequency range. Among the reported vibrational energy harvesters using electromagnetic [5]–[8], electrostatic [9]–[11] and piezoelectric [12]–[15] energy conversion transducers, most if not all of the devices have been characterized by using swept-sine vibration shaker at the laboratory. Thus they may focus on the power generation and vibration behavior of linear oscillators in response to harmonic excitations without considering the irregular vibration frequencies occurring in practical applications.

Nevertheless, some research groups have made the attempts to implement vibrational energy harvesters to the real environment or conducted their measurement under condition close to potential applications. For example, the harvester devices have been tested on engine block of a car [16], different bridge locations [17], human shoes [18] and pocket [19], [20]. Comparing to vibration sources induced by machines and buildings, energy harvesting from human motions are particularly challenging due to the low frequency range (<10 Hz), and random or even unpredictable vibration pattern. As a result, the most common linear mass-spring-damper systems with narrow frequency range are poorly suitable for energy harvesting from human motions because the output power of a linear harvester drops dramatically under off-resonance conditions [21]. This problem can be overcome by using frequency wideband mechanisms, such as harvester array [22], [23], mechanical stoppers [24]-[26], nonlinear springs [27], [28], frequency tunable mechanisms [29]-[32] and multi-frequency harvesters [33]-[36]. However, most of the reported wideband harvesters operate at relatively high frequencies. In order to achieve low resonant frequency to match with the vibration frequency associated with human motions, i.e. less than 10 Hz, a very compliant spring structure, which is not only the major root cause of damage, but also requires enough space to permit large mechanical displacement.

Saha et al [37] have successfully demonstrated the possibility of harvesting energy from human walking and slow running by using a nonlinear electromagnetic generator with magneticspring. Such type of nonlinear energy harvester with magneticspring offers many advantages of high output performance, high tunability, less prone to failure, ease of construction, and low cost. Hence, similar prototypes have been reported by several groups from different aspects. A theoretical analysis of the nonlinear Duffing oscillator using magnetic levitation has been investigated by Mann and Sims [38]. Dallago et al [39] have taken into account nonlinear effect and built an

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Fig. 1. Schematic diagram of the nonlinear EMEH device.

analytical model of the magnetic-spring based energy harvester. Experimental results have validated that the model is able to predict the voltage performance of the harvester. Foisal et al [40] have presented an array of harvesters with magneticspring and demonstrated the possibility of harvesting energy from multiple frequencies. A similar EMEH using multi-pole magnet has also been studied recently [41] by the same group. On top of these works, it is quite necessary to proceed an optimization of the key parameters such that a feasible design methodology of the magnetic-spring based nonlinear energy harvester could gain more impressive output performance. In this work, we fabricated nonlinear EMEH with various dimension parameters in terms of tube length and winding coil width. The comparison and optimization has been performed based on the modeling and hand shaking test which is aiming at practical applications.

II. DESIGN AND MODELING

A. Prototype Structure

A schematic diagram of a nonlinear EMEH with magneticspring is shown in Fig. 1. It is composed of a cylindrical tube made of acrylic glass, two fixed disc magnets of 3-mmdiameter and 2-mm-height, a center moving disc magnet of 6-mm-diameter and 6-mm-height and copper induction coils. The left and right disc magnets are attached respectively on the left and right caps. The three disc magnets are placed in such a way that the same magnetic poles face to each other. Thus the magnetic repulsive force from the left and right magnets serves as magnetic spring and levitates the center disc magnet in the middle of the tube. The copper coils are wrapped at the middle of the outside of the tube with a total coil length l_c and winding coil width C_w . According to Faraday's law of induction, the external ambient vibration causes the center moving magnet to oscillate, which in turn induces change in flux over the winding coil thereby generating electric current in copper coils. To reduce the air damping during the movement of the magnetic mass inside the tube, four distributed throughholes of 3-mm in diameter are drilled on both the left and right caps.

In this work, at the beginning nine different devices are fabricated with three different tube lengths of 46, 56

TABLE I PARAMETERS OF THE NONLINEAR EMEH DEVICE

Symbol		
Magnetic material	NdFeB (N35)	
Coil material	Copper	
Housing material	Acrylic	
Young's modulus (Pa)	1.53×10^{11}	
Coercive Force (kA/m)	860	
Residual magnetic flux density (Br) (T)	1.18	
Diameter and height of left magnet (d_L, h_L) (mm)	3×2	
Diameter and height of right magnet (d_R, h_R) (mm)	3×2	
Diameter and height of center magnet (d_M, h_M) (mm)	6×6	
Mass of the moving magnet (inertial mass) (m) (g)	1.23	
Coil resistance $(R_{int})(\Omega)$	10, 15, 20	
Wire diameter (d_c) (mm)	0.2	
Total wire length (l_c) (m)	20	
Coil width (C_W) (mm)	10, 20, 30	
Tube length (l_T) (mm)	46, 56, 66	
Volume (cm ³)	4.37, 5.32, 6.27	
Inner radius (r_{in}) (mm)	3.5	
Outer radius (r_{out}) (mm)	5.5	
Spacing between moving magnet and $coil(G)(mm)$	2.5	
Magnetic flux density (B) (T)	0.1786	

and 66 mm, respectively. Each of them has the same total coil length l_c of 20 m but three different widths of winding coils C_w of 10, 20 and 30 mm, respectively. The output performances of these devices are compared to obtain the optimized device configuration. Then, by changing two important parameters, i.e. total wired coil length and the coil width, the output performance of the optimized device is improved further. The parameters of the EMEH devices are listed in Table 1 and the device optimization is discussed in later section.

B. Analytical Modeling

A vibration based energy harvester is essentially a secondorder spring-mass-damper system [21]. The first reference frame in Fig. 1, designated as z, describes the motion of tube housing when the input excitation $z(t) = Acos\Omega t$ is applied in which, A and Ω represent the motion amplitude and the excitation frequency, respectively. The second reference frame, designated as x, represents the motion of center moving magnet. As the model describes in-plane movement, we neglect the acceleration due to gravity g. According to Newton's second law, the equation of motion can be written as [38]:

$$m\ddot{x} + c_m(\dot{x} - \dot{z}) + k(x - z) + k_3(x - z)^3 + \alpha i = 0 \quad (1)$$

where *m* is the mass of moving magnet; c_m is mechanical damping coefficient; *k* is linear spring constant; k_3 is nonlinear spring constant; *x* is the amplitude of center magnet; *z* is the amplitude of tube housing; *i* is electric current; $\alpha = NBl$ is electromagnetic coupling coefficient, where *N* is number of coil turns and *l* is the effective coil length; B is the magnetic flux density of a cylindrical magnet, and is given as [42]:

$$B = \frac{B_r}{2} \left(\frac{(G+h_M)}{\sqrt{(d_M/2)^2 + (G+h_M)^2}} - \frac{G}{\sqrt{(d_M/2)^2 + G^2}} \right) \quad (2)$$



Fig. 2. (a) Circuit diagram of the EMEH device. (b) Magnetic force illustration of the three disc magnets.

where B_r is the residual magnetic flux, G is the spacing between the moving magnet and coil, and h_M and d_M are the height and diameter of the moving mass, respectively.

By applying Kirchhoff's law to the electric circuit in Fig. 2(a), the expression for electric current neglecting coil inductance can be obtained as:

$$i = \alpha \frac{\dot{x} - \dot{z}}{R_{load} + R_{int}} \tag{3}$$

where R_{load} and R_{int} are the load and internal resistances of the device, respectively. By substituting the value of current from Eq. (3) into Eq. (1), we have:

$$m\ddot{x} + c(\dot{x} - \dot{z}) + k(x - z) + k_3(x - z)^3 = 0$$
(4)

where $c = c_m + c_e$ is the total damping coefficient, comprising both mechanical and electrical damping coefficient. The electrical damping coefficient is given by $c_e = \frac{\alpha^2}{R_{load} + R_{int}}$. As show in Fig. 2(b), the magnetic force between left fixed

As show in Fig. 2(b), the magnetic force between left fixed magnet and right moving magnet is given as [3]:

$$F_L = \frac{\mu_0 Q_L Q_M}{4\pi r_1^2}$$
(5)

where r_1 is the distance between left and center magnet poles; the magnetic field intensity is $Q = H_cA$, H_c is the coercive force, and A is the pole surface area. Similarly, the magnetic force between right fixed magnet and center moving magnet is given as:

$$F_R = \frac{\mu_0 Q_R Q_M}{4\pi r_2^2} \tag{6}$$

where r_2 is the distance between right and center magnet poles. For the case of in-plane movement of the tube housing, we have the expression for r_1 and r_2 as:

$$r_1 = r_2 = \frac{l_L - (h_L + h_R + h_M)}{2} \tag{7}$$

As the center moving magnet is denoted by distance x as shown in Fig. 2(b), the resultant magnetic force applied to the center moving magnet is obtained as:

$$F_{res} = F_L - F_R = \frac{\mu_0 Q_M}{4\pi} \left(\frac{Q_L}{(r_1 + x)^2} - \frac{Q_R}{(r_2 - x)^2}\right) \quad (8)$$

where F_{res} also represents the magnetic spring restoring force, and is given as:

$$F_{res} = kx + k_3 x^3 \tag{9}$$

where k and k_3 are the linear and nonlinear spring constants, respectively.

By considering the relative displacement between moving magnet and tube housing, we introduce y = x - z and modify the governing equation of motion from Eq. (4) as:

$$\ddot{y} + 2\zeta \omega m \dot{y} + \omega^2 y + \beta y^3 = F_e cos \Omega t$$
(10)

where the damping ratio is $\zeta = \frac{c}{2\omega m}$, the resonant frequency is $\omega = 2\pi f_0 = \sqrt{\frac{k}{m}}$, the excitation force per unit mass is $F_e = \Omega^2 A$ and the ratio between nonlinear spring constant and mass of moving magnet is $\beta = \frac{k_3}{m}$. The modified frequency response equation by solving Eq. (10) is given as:

$$(\frac{3\beta}{8\omega})^2 a^6 + \left(\frac{3}{4}\beta\left(1-\frac{\Omega}{\omega}\right)\right)a^4 + ((\Omega-\omega)^2 + (\zeta\omega)^2)a^2 - (\frac{F_e}{2\omega})^2 = 0 \quad (11)$$

The detailed derivation can be found in ref [38]. Eq. (11) gives the frequency response amplitude for the relative displacement of the center moving magnet (|y| = a) against the excitation frequency (Ω), where the relative displacement is given by $y = a\cos(\Omega t - \Theta)$. Hence, the predicted relative velocity $\dot{y} = -\Omega a \sin(\Omega t - \Theta)$ can be obtained, which is more meaningful for the calculation of coil induced voltage as:

$$|V_{\rm oc}| = \alpha |\dot{y}| \tag{12}$$

By applying Kirchhoff's law to circuit resented in Fig. 2(a) we have:

$$i = \frac{V_{oc}}{(R_{load} + R_{int})} \tag{13}$$

Hence, the maximum averaged power delivered to the load is given by:

$$P_{avgmax} = \left(\frac{\alpha \Omega_{max} a_{max}}{R_{load} + R_{int}}\right)^2 R_{load} \tag{14}$$

where a_{max} is the maximum amplitude of the frequency response when the system is excited at $\Omega = \Omega_{max}$. The relationship between them is obtained by setting $\frac{da}{d\Omega} = 0$ in Eq. (11) as:

$$\Omega_{max} = \frac{3\beta}{8\omega} a_{max}^2 + \omega \tag{15}$$

The expression for a_{max} is obtained by substituting $\Omega = \Omega_{max}$ in Eq. (11) as:

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$$a_{max} = \frac{F_e}{2\zeta\,\omega^2}\tag{16}$$

By substituting a_{max} in Eq. (16) into Eq. (15), we have:

$$\Omega_{max} = \frac{3\beta F_e^2}{32\omega^5 \zeta^2} + \omega \tag{17}$$



Fig. 3. Simulation of the force-displacement curves for linear region of the tube lengths of: (a) 46 mm, (b) 56 mm, (c) 66 mm and nonlinear region of the tube lengths of (d) 46 mm, (e) 56 mm, and (f) 66 mm.

TABLE II SIMULATED SPRING CONSTANT OF THE DEVICE FOR DIFFERENT TUBE LENGTHS

Tube length (l_T) (mm)	Linear spring constant (k) (N/m)	Nonlinear spring constant (k_3) (N/m^3)	Resonant frequency (f_0) (Hz)
46	10	$7.6 imes 10^4$	14.36
56	4.9	$2.1 imes 10^4$	10.05
66	2.7	$7.4 imes 10^3$	7.46

Therefore, by substituting the expressions of Ω_{max} and a_{max} into Eq. (15), the maximum averaged power can be expressed as:

$$P_{avgmax} = (\alpha F_e \frac{3F_e^2\beta + 32\zeta^2\omega^2}{64\zeta^3\omega^7(R_{int} + R_{load})})^2 R_{load}$$
(18)

C. Simulation

For the nonlinear energy harvester with magnetic spring, the resultant force applied to the center moving magnet is expressed by Eq. (8), by which the force-displacement relationship can be obtained. For different tube lengths of 46, 56 and 66 mm, the magnetic resultant forces against displacements are shown in Fig. 3. By curve fitting in MATLAB, the linear and nonlinear spring constants for different tube lengths can be estimated from the linear region (-2 to 2 mm) in Figs. 3(a)-(c) and the nonlinear region (-6 to 6 mm) in Figs. 3(d)-(f). The resonant frequency can be obtained from the linear spring constant. The estimated parameters are listed in Table 2. In the case that the tube length increases, the linear spring constant as well as the resonant frequency decreases. Thus it is preferable condition for the harvesting energy from hand shaking of low frequency vibrations.

The open circuit voltage response can be obtained by executing Eqs. (11) and (12) using MATLAB. The device



Fig. 4. Simulation of open circuit voltage response of the optimized EMEH device for applied accelerations of: (a) 0.5g; (b) 0.55g; and (c) 0.65g.



Fig. 5. (a) Schematics diagram of handshaking test setup. (b) Photograph of the EMEH device for in-plane handshaking test.

with tube length of 66 mm and winding coil width of 10 mm are used in the simulation and the damping ratio (ζ) is assumed to be 0.1. Fig. 4 shows the open circuit voltage against excitation frequency at different accelerations of 0.5g, 0.55g and 0.65g, where the nonlinear response effect due to the nonlinear magnetic spring force is observed. The open circuit voltages reach their maxima of 1.89, 2.09 and 2.64 V at the shifted resonant frequencies of 8.4, 8.6 and 9.0 Hz, respectively for the excitation accelerations of 0.5g, 0.55g and 0.65g, respectively. From this point of view, the nonlinearity of the EMEH with magnetic-spring is strengthened as the input acceleration increases.

III. EXPERIMENTS AND DISCUSSION

The experimental setup for handshaking test of the EMEH device is shown in Fig. 5. The acceleration magnitude and frequency of the hand shaking is measured by a small and low power accelerometer from Analog Device (ADXL325). A data acquisition (DAQ) card of NI USB-6525 is used to collect signal outputs from the harvester device and accelerometer. The data collected from the DAQ card is processed and displayed in the LabView software. The signal output of the accelerometer is then processed in Matlab by performing FFT (Fast Fourier transform) to convert into frequency domain thereby obtaining value of applied acceleration and frequency.



Fig. 6. (a) Time domain and (b) frequency domain characteristics for tube.

A. Device Optimization

Nine different devices with three tube lengths of 46, 56 and 66 mm, while each with winding coil widths of 10, 20 and 30 mm, are prepared. There are two categories of comparison in the experiment: (a) varying winding coil width while keeping the tube length constant; and (b) varying tube length while keeping the winding coil width constant. To ensure the comparability and stability of the results, all the nine devices are tested by in-plane hand shaking of about 1 cm displacement. In the experiment, a relatively constant acceleration of 0.45-0.6 g and shaking frequency of 4.5-4.8 Hz are maintained. A load resistance of 10 Ω which matches with the internal coil resistance is connected to gain the maximum power transfer. The time domain characteristics depicted in Fig.6 consists of both EMEH device and accelerometer characteristics obtained from DAQ card. The output root mean square (RMS) voltage is obtained from time domain characteristics of EMEH device and hence we obtain output power delivered to load. Fig. 6 shows the time domain and frequency domain characteristics for tube length of 66 mm and coil width of 10 mm. Thus we obtain the output for nine different device configurations listed in Table 3 and the comparison for output



Fig. 7. Comparison of the nine nonlinear EMEH devices with various tube lengths and winding coil widths.



Fig. 8. RMS voltage and output power against load resistance of the optimized device.



Fig. 9. Output power against excitation frequency of the optimized device for in-plane displacements of 1 and 2 cm.

power delivered to the load for different device configuration is shown in Fig. 7.

For the first comparison with a constant tube length of 46 mm, as the winding coil width increases from 10, 20 to 30 mm, the power delivered to the load decreases from 115.9, 63.25 to 15.4 μ W, respectively. The similar trends have been observed for the tube lengths of 56 and 66 mm. This is because when the magnet enters the winding coil, there is an increasing magnetic flux linkage which produces positive voltage through the coil. When the magnet is completely enclosed by the winding coil, the total flux linkage through the coil is

 TABLE III

 Comparison of the Output Performance of Nine Nonlinear EMEH Devices

Tube length	Coil width	Shaking	Shaking frequency	RMS voltage	Power	Power density
(mm)	(mm)	acceleration (g)	(Hz)	(mV)	(µW)	(µW /cm ³)
46	10	0.59	4.5	34.05	115.9	26.51
	20	0.59	4.8	25.15	63.25	14.47
	30	0.6	4.6	12.41	15.4	3.52
56	10	0.45	4.4	47.41	224.77	42.25
	20	0.6	4.5	37.52	140.77	26.45
	30	0.45	4.5	23.23	53.96	10.14
66	10	0.54	4.5	53.13	284.28	45.34
	20	0.52	4.7	44.16	194.81	31.06
	30	0.46	4.5	32.79	107.52	17.14

TABLE IV

COMPARISON OF THE SIMULATED AND EXPERIMENTAL RESULTS OF THE OPTIMIZED EMEH DEVICE

Shaking acceleration (g)		Shaking frequency (Hz)		Open circuit voltage (V)	
Simulated	Experimental	Simulated	Experimental	Simulated	Experimental
0.5	0.51	4.2	4.2	0.20	0.25
0.55	0.56	4.6	4.7	0.28	0.27
0.65	0.65	4.8	4.8	0.36	0.30
0.65	0.65	4.8	4.8	0.36	0.30



Fig. 10. Time domain voltage waveforms of the optimized device for the in-plane displacement of 1cm (a) and 2 cm (b) at hand shaking frequency of 6.7 Hz.

constant and thus the voltage induced becomes zero. As the magnet leaves the winding coil, there is a decreasing magnetic flux linkage, and the induced voltage is then negative. Clearly, in order to maximize the output voltage using the same total wired length of 20 m, the winding width of the coil should be limited so that the magnetic flux is not completely enclosed by the coil over any significant part of the generation period [43]. Therefore, the EMEH device with shorter winding coil width of 10 mm performs better and delivers higher power output to the load than the longer winding coil widths of 20 and 30 mm for the same tube length.

The second comparison considers a constant winding coil width but varying tube length. It is seen in Fig. 7 that at the winding coil width of 10 mm, the power delivered to the load are 115.9, 224.7 and 284.28 μ W for tube lengths of 46, 56 and 66 mm, respectively. The same holds good for winding coil widths of 20 and 30 mm. This is because the linear spring constant decreases with increasing tube length. For a longer tube, the magnetic repulsive force decreases and the inertial

mass has a displacement of greater amplitude. For the same external excitation, the displacement will take longer time resulting in lower movement frequency. Hence, the resonant frequency of the system decreases (as described in Table 2) and gets closer to the desired handshaking frequency of about 5 Hz. In addition, larger movement of the inertial mass also implies greater variation of the magnetic field acting on the coil. Thus, the device with longer tube length performs better and delivers higher power to the load. Based on the above comparison, it can be conclude that the optimized configuration among the nine EMEH devices possesses the longest tube length of 66 mm and the shortest coil width of 10 mm. Hence, the optimized device with the tube length of 66 mm is selected in the following in-plane hand shaking test.

B. Output Voltage and Power of Optimized Device

The open circuit voltages of the optimized device generated by hand shaking excited at accelerations of 0.51, 0.56 and 0.65g and frequencies of 4.2, 4.7 and 4.8 Hz are 0.25, 0.27 and 0.3 V, respectively. The corresponding simulated open circuit voltages denoted as points A', B', and C' in Fig. 4 are 0.2, 0.28 and 0.36 V, respectively, at corresponding excitation accelerations of 0.5, 0.55 and 0.65g and frequencies of 4.2, 4.6 and 4.8 Hz. Table 4 shows a comparison between the tested and simulated open circuit voltages. As can be seen, the experimental and modeling results are quite comparable. Figure 8 shows the RMS voltage and power output of the optimized device for different load resistances obtained from handshaking with in-plane displacement of 1 cm and excitation frequency of 4.5 Hz. The maximum power of 311.7 μ W is obtained when the load resistance matches with the internal coil resistance of 10 Ω .

Figure 9 shows the output power against different excitation frequencies from 2.7 to 6.7 Hz with respect to the in-plane hand shaking displacements of 1 and 2 cm, respectively, at matched load resistance of 10 Ω . As one expected, the output power increases as the excitation acceleration and frequency increase under the same displacement amplitude. The maximum power obtained from the in-plane displacements of 1 and 2 cm are 568.66 and 652.38 μ W, respectively, at the handshaking limit frequency of 6.7 Hz. The corresponding power densities obtained are 90.67 and 104.02 μ W/cm³, respectively. The time-domain voltage waveform at corresponding maxima's of displacement 1 and 2 cm are shown in Fig. 10.

IV. CONCLUDING REMARKS

In this work, we have designed and fabricated EMEH devices based on nonlinear magnetic-spring configuration. The devices with parameters such as tube length of 46, 56 and 66 mm and winding coil width of 10, 20 and 30 mm are investigated via simulation and hand shaking experiment. It is found that a longer tube length reduces the operating frequency of the nonlinear device, and hence more power can be generated when the resonant frequency approaches the driving handshaking frequency. Meanwhile, a limited winding coil width should be considered such that the magnetic flux is not completely enclosed by the coil over the generation period and thus produce more power. In order to make the nonlinear EMEH device more applicable to harvesting irregular human motion of less than 5 Hz, the resonant frequency will be lowered in further study by reducing the magnetic force and optimizing the fixed and movable magnets. The main advantage of the proposed nonlinear EMEH device is its relatively large power and power density of 654.38 μ W and 104.02 μ W/cm³, respectively, produced from a very low hand shaking frequency of less than 10 Hz. Therefore, with further improvements, it could be potentially applicable for the self-powered wearable health monitoring devices, such as smart watch.

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